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APPLICATION NOTE

ON

RESISTANCE-COUPLED AUDIO-FREQUENCY AMPLIFIERS

Difficulties from low-frequency oscillation (motorboating) are often experienced when a high-gain multi-stage audio amplifier of conventional design obtains B-supply voltage from a single power-supply unit. These difficulties are usually due to interstage coupling through a common impedance in the power unit. Customary expedients to prevent motorboating include the use of very large filter condensers and several power-supply units. However, by suitable design, the gain of an audio amplifier of low frequencies can be reduced to such levels that the effects of feed-back through a power unit of conventional design are greatly reduced.

The use of a series screen resistor and a self-bias resistor offers several advantages over fixed-voltage operation: (1) the effects of possible tube differences are minimized; (2) operation over a wide range of plate-supply voltages without appreciable change in gain is feasible; and (3) the low frequency at which the amplifier cuts off is easily changed. Fixed-bias or fixed-screen-voltage operation increases the tendency of an amplifier to motorboat and decreases the compensating action of the remaining series resistors. The advantages of an amplifier constructed according to the data presented in this Note can be further emphasized by the addition of suitable decoupling resistors and condensers. With a proper decoupling filter in the plate circuit of each stage, three or more amplifier stages can be operated from a single power-supply unit of conventional design without encountering any difficulties due to coupling through the power unit; not more than two stages should be operated from a single power-supply unit when decoupling filters are not used.

Detailed information on the operation of the 2B7, 6B7, 6B8, 6C6, 6J7, and 57 as resistance-coupled audio-frequency amplifiers is presented in the accompanying Pentode Chart. These data obtain for plate-supply voltages from 45 to 600 volts, for plate-resistor values of 0.1, 0.25, and 0.5 megohm, and for a number of grid-resistor values. The combination of resistor and condenser values that is suggested in the Pentode Chart causes a 30 per cent drop in output voltage per stage at 100 cycles. A similar cut-off characteristic at any other low frequency (f_1) can be obtained by multiplying the capacitance values shown in the chart by $100/f_1$.

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The values of resistors and condensers suggested in the Pentode Chart were determined as follows. A cathode-ray oscillograph was connected to show the relation between output voltage and input voltage (dynamic characteristic) of a single stage resistance-coupled amplifier. From previous experience, it is known that in an a-f amplifier employing pentode-type tubes the output voltage is substantially independent of frequency above the cut-off frequency (f_1) and below some higher frequency (f_2). It is also known that the value of the parallel combination of R_L and R_g determines to some extent the value of (f_2). The following approximate values of (f_2) obtain for the recommended values of R_L when normal circuit and tube capacitances are present.

R_L^*	f_2
0.1 megohm	20,000 cycles
0.25 "	10,000 cycles
0.5 "	5,000 cycles.

*For the condition where R_L is less than R_g .

For given values of R_L and R_g , the magnitude of R_o , R_d , and the input signal were adjusted until kinks just appeared at the extremities of the dynamic characteristic at the grid-current point. Under these conditions, the tube operates at the center of its dynamic characteristic, which is nearly linear throughout the operating range. Therefore, maximum output at minimum distortion is obtained. The values C_o , C_d , and C were then determined. Very large capacitances were used in the positions indicated by C_o , C_d , and C . The value of each was reduced, in turn, until the output voltage dropped 10 per cent at 100 cycles. Hence, for each stage of amplification, the output voltage at 100 cycles is approximately $0.7 E_o$, where E_o is the output voltage at some higher frequency which is less than f_2 . For (n) like stages in cascade, the output voltage at 100 cycles is $(0.7 E_o)^n$. For a given value of decoupling resistor, the size of the decoupling condenser can be determined in the manner similar to that used for obtaining the values of C_o , C_d , and C .

Detailed information on the operation of triodes as resistance-coupled audio-frequency amplifiers is presented in the accompanying Triode Chart. The combination of condenser and resistor values suggested in this Triode Chart causes a 20 per cent drop in output voltage per stage at 100 cycles. A similar cut-off characteristic at any other low frequency (f_1) can be obtained by multiplying the capacitance values shown by $100/f_1$. As with pentodes, the use of self-bias reduces the effects of possible tube differences and permits operation over a wide range of plate-supply voltages without appreciable change in gain. The regulating action of a self-biased triode amplifier is not as good as that of a pentode amplifier having series screen and self-bias resistors, because the regulating action of a screen is not available in a triode.

When a number of high-mu triode amplifier stages are cascaded, the high-frequency response may be severely curtailed, because the high effective input capacitance of a triode shunts the load of the previous tube.

When good high-frequency response from a triode amplifier is desired, therefore, low- μ tubes and low values of plate and grid resistors should be used.

On the Charts, the values of C_o are given for an amplifier with d-c heater excitation. When a-c is used, depending on the character of the associated circuits, the gain, and the value of f_1 , it may be necessary to increase the value of C_o to minimize hum disturbances. It may also be desirable to have a d-c potential difference of approximately 10 volts between heater and cathode.

RESISTANCE-COUPLED AMPLIFIER CHART FOR PENTODES

	90				180			
	0.1	0.25	0.5	1	0.1	0.25	0.5	1
PLATE-SUPPLY VOLTAGE (E _{bb}) - Volts ¹	0.1	0.25	0.5	1	0.1	0.25	0.5	1
GRID RESISTOR (R _g) - Megohms ²	0.37	0.44	0.44	1.1	0.44	0.53	0.53	1.1
SCREEN RESISTOR (R _d) - Megohms	1.20	1.10	1.30	2.40	1.00	1.10	1.10	2.00
CATHODE RESISTOR (R _c) - Ohms	0.05	0.05	0.05	0.05	0.05	0.05	0.05	0.05
SCREEN BY-PASS CONDENSER (C _d) - μf	5.2	5.3	4.8	3.7	6.5	6.7	6.7	5.2
CATHODE BY-PASS CONDENSER (C _c) - μf	0.02	0.01	0.006	0.008	0.02	0.01	0.006	0.008
BLOCKING CONDENSER (C) - μf	17	22	33	23	42	52	59	45
VOLTAGE OUTPUT (E _o) - Peak Volts ³	41	55	66	70	51	69	83	135
VOLTAGE GAIN ⁴								

PENTODE TYPES: 6C6, 6J7, 57

	300				600 (max.)			
	0.1	0.25	0.5	1	0.1	0.25	0.5	1
PLATE-SUPPLY VOLTAGE (E _{bb}) - Volts ¹	0.1	0.25	0.5	1	0.1	0.25	0.5	1
GRID RESISTOR (R _g) - Megohms ²	0.44	0.5	0.53	1.18	0.48	0.53	0.53	1.18
SCREEN RESISTOR (R _d) - Megohms	500	450	600	1100	200	250	300	500
CATHODE RESISTOR (R _c) - Ohms	0.07	0.07	0.06	0.04	0.09	0.08	0.08	0.05
SCREEN BY-PASS CONDENSER (C _d) - μf	8.5	8.3	8	5.5	11.5	11.1	10.5	8.2
CATHODE BY-PASS CONDENSER (C _c) - μf	0.02	0.01	0.006	0.008	0.02	0.01	0.006	0.008
BLOCKING CONDENSER (C) - μf	55	81	96	81	90	140	150	125
VOLTAGE OUTPUT (E _o) - Peak Volts ³	61	82	94	104	77	100	112	136
VOLTAGE GAIN ⁴								

	90				180			
	0.1	0.25	0.5	1	0.1	0.25	0.5	1
PLATE-SUPPLY VOLTAGE (E _{bb}) - Volts ¹	0.1	0.25	0.5	1	0.1	0.25	0.5	1
GRID RESISTOR (R _g) - Megohms ²	0.37	0.5	0.6	1.18	0.44	0.5	0.6	1.18
SCREEN RESISTOR (R _d) - Megohms	2000	2200	2000	3500	1000	1200	1200	1900
CATHODE RESISTOR (R _c) - Ohms	0.07	0.07	0.06	0.04	0.08	0.08	0.07	0.05
SCREEN BY-PASS CONDENSER (C _d) - μf	3	3	2.8	1.9	4.4	4.4	4	2.7
CATHODE BY-PASS CONDENSER (C _c) - μf	0.02	0.01	0.006	0.008	0.02	0.015	0.008	0.01
BLOCKING CONDENSER (C) - μf	19	28	29	26	30	52	53	39
VOLTAGE OUTPUT (E _o) - Peak Volts ³	24	33	37	43	30	41	46	55
VOLTAGE GAIN ⁴								

DUPLEX-DIODE PENTODE TYPES: 2B7, 6B7, 6B8

	300				600 (max.)			
	0.1	0.25	0.5	1	0.1	0.25	0.5	1
PLATE-SUPPLY VOLTAGE (E _{bb}) - Volts ¹	0.1	0.25 <td>0.5</td> <td>1</td> <td>0.1</td> <td>0.25 <td>0.5</td> <td>1</td> </td>	0.5	1	0.1	0.25 <td>0.5</td> <td>1</td>	0.5	1
GRID RESISTOR (R _g) - Megohms ²	0.1	0.25 <td>0.5</td> <td>1</td> <td>0.1</td> <td>0.25 <td>0.5</td> <td>1</td> </td>	0.5	1	0.1	0.25 <td>0.5</td> <td>1</td>	0.5	1
SCREEN RESISTOR (R _d) - Megohms	950	1100	900	1500	350	500	350	1000
CATHODE RESISTOR (R _c) - Ohms	0.09	0.09	0.08	0.06	0.1	0.1	0.1	0.08
SCREEN BY-PASS CONDENSER (C _d) - μf	4.6	5	4.8	3.2	6.6	7	7	4
CATHODE BY-PASS CONDENSER (C _c) - μf	0.025	0.015	0.009	0.015	0.03	0.02	0.01	0.015
BLOCKING CONDENSER (C) - μf	60	89	86	70	72	130	120	135
VOLTAGE OUTPUT (E _o) - Peak Volts ³	36	47	54	64	43	59	67	76
VOLTAGE GAIN ⁴								

¹ Voltage at plate equals Plate-Supply Voltage minus voltage drop in R_L and R_c. For other supply voltages differing by as much as 50% from those listed, the values of resistors, condensers, and gain are approximately correct. The value of voltage output, however, for any of these other supply voltages equals the listed voltage output multiplied by the new plate-supply voltage divided by the plate-supply voltage corresponding to the listed voltage output. The 600-volt conditions are maximum.

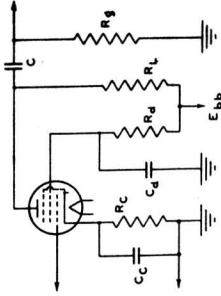
² For following stage (see Circuit Diagram).

³ Voltage across R_g at grid-current point.

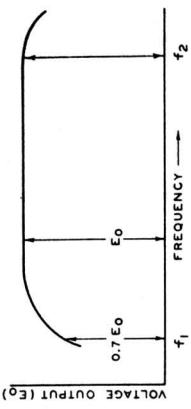
⁴ At 5 volts (RMS) output. Gain at full output is somewhat less than the listed values.

EXPLANATION

DIAGRAM WITH LEGEND



FREQUENCY CHARACTERISTIC OF SINGLE-STAGE RESISTANCE-COUPLED PENTODE AMPLIFIER



REMARKS

- A. Condensers C_c, C_d, and C_b have been chosen to give output voltages equal to 0.7 E_o for f₁ of 100 cycles. For any other value of f₁, multiply values of C_c, C_d, and C_b by 100/f₁. In the case of condenser C_b, the values shown are for an amplifier with d-c heater excitation. When a-c is used, depending on the character of the associated circuits, the gain, and the value of f₁, it may be necessary to increase the value of C_b to maintain the response at approximately 10 volts between heater and cathode.
- B. f₂ = frequency at which high-frequency response begins to fall off.
- C. The voltage output at f₁ for π like stages equals 10.7 E_o².
- D. Decoupling filters are not necessary for two stages or less.
- E. Approximate values of f₂ for different values of R_L for an amplifier of typical construction are as follows: R_L = 0.1 meg., 20000 cps; R_L = 0.25 meg., 10000 cps; R_L = 0.5 meg., 5000 cps. Use of R_g.
- F. A variation of ±10% in values of resistors and condensers has only slight effect on performance.

RESISTANCE - COUPLED AMPLIFIER CHART FOR TRIODES

R_d = SCREEN RESISTOR (Megohms) R_L = PLATE RESISTOR (Megohms)
 R_g = GRID RESISTOR (Megohms) V_G = VOLTAGE GAIN

C_d = BLOCKING CONDENSER (μf) E_o = VOLTAGE OUTPUT (Peak Volts)
 C_c = CATHODE BY-PASS CONDENSER (μf) R_c = CATHODE RESISTOR (Ohms)

TRIODE TYPES : 2A6,75

Ebb 1	90					180					300					Ebb 1				
	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2		0.25	0.5	1	2
Rg 2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.25	0.5	1	2	
Rc	6300	6600	6700	10000	11500	2600	2900	3000	4300	4800	5300	8000	8800	1900	2200	2300	3300	3900	4200	7000
Cc	2.2	1.7	1.7	1.24	1.07	3.3	2.9	2.7	2.1	1.8	1.5	1.1	0.9	4	3.5	2.7	2	1.8	1.6	
C	0.02	0.01	0.006	0.01	0.006	0.025	0.015	0.007	0.015	0.007	0.004	0.007	0.002	0.03	0.015	0.007	0.015	0.007	0.004	
Eo 3	3	5	6	5	7	16	22	23	21	28	33	25	38	31	41	45	42	51	60	
V.G. 4	25 d	29 b	31 c	34 b	40 c	29	36	37	43	53	52	57	58	31	39	42	48	53	56	

TRIODE TYPE 6F5

Ebb 1	90					180					300					Ebb 1			
	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2		0.25	0.5	1
Rg 2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.25	0.5	1	2
Rc	4400	4800	5000	8000	9000	1800	2000	2200	3500	4100	4500	6900	7700	1300	1600	1700	2600	3200	3500
Cc	2.5	2.1	1.8	1.33	1.18	4.4	3.3	2.9	2.3	1.7	1.3	0.9	0.83	5	3.7	3.2	2.5	2.1	1.5
C	0.02	0.01	0.005	0.01	0.005	0.025	0.015	0.006	0.01	0.006	0.004	0.006	0.0015	0.025	0.01	0.006	0.01	0.007	0.004
Eo 3	4	5	6	6	7	16	23	25	21	26	32	24	37	33	43	48	41	54	63
V.G. 4	28 d	34 b	35 c	39 b	43 c	37	44	46	48	53	57	53	66	42	49	52	56	63	67

TWIN-TRIODE TYPES : 6A6, 6N7, 53 (ONE TRIODE UNIT)

Ebb 1	90					180					300					Ebb 1			
	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2		0.25	0.5	1
Rg 2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.25	0.5	1	2
Rc	1900	2250	2500	4050	4950	1300	1700	1950	2950	3800	4300	6600	7650	1150	1500	1750	2650	3400	4000
Cc	0.025	0.01	0.006	0.01	0.006	0.03	0.015	0.007	0.015	0.007	0.0035	0.002	0.03	0.015	0.007	0.015	0.0055	0.003	0.0015
Eo 3	13	19	20	16	20	35	46	50	40	50	57	44	61	60	83	86	75	87	100
V.G. 4	16	19	20	20	22	19	21	22	23	24	24	25	25	20	22	23	23	24	24

TWIN-TRIODE TYPE 79 (ONE TRIODE UNIT)

Ebb 1	90					180					300					Ebb 1			
	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2		0.25	0.5	1
Rg 2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.25	0.5	1	2
Rc	2050	2200	2350	4000	4250	1050	1250	1350	2050	2450	2750	4100	4650	800	1000	1100	1650	2050	2350
Cc	0.04	0.015	0.009	0.015	0.006	0.04	0.02	0.009	0.02	0.01	0.005	0.0035	0.002	0.025	0.01	0.006	0.01	0.0055	0.003
Eo 3	5.8	8.4	9.5	7.1	9.7	21	27	31	26	34	40	39	44	40	57	60	56	66	77
V.G. 4	23 b	29 c	29	31 c	35	27	31	34	37	41	42	44	45	29	34	36	39	42	43

TRIODE TYPES : 56, 76

Ebb 1	90					180					300					Ebb 1			
	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1		0.25	0.5	1
Rg 2	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	0.1	0.25	0.5	1
Rc	2500	3200	3800	4500	6500	2400	3000	3700	4500	6500	7600	10700	17700	2400	3100	3800	4500	6400	7500
Cc	2	1.6	1.25	1.05	0.82	2.5	1.9	1.65	1.45	0.97	0.8	0.6	0.4	2.8	2.2	1.8	1.6	1.2	0.98
C	0.06	0.03	0.015	0.03	0.015	0.06	0.035	0.015	0.035	0.015	0.008	0.015	0.007	0.08	0.045	0.02	0.04	0.02	0.009
Eo 3	16	21	23	19	23	36	48	55	45	55	57	49	64	65	80	95	74	95	104
V.G. 4	7	7.7	8.1	8.1	8.9	7.7	8.2	9	9.3	9.5	9.8	9.7	10	8.3	8.9	9.4	9.5	10	10

(continued from preceding page)

TYPES: 6C5 (TRIODE), AND 6C6, 6J7, 57 (AS TRIODES)

Ebb 1	90					180					300					Ebb 1
	RL	Rg 2	Rc	Cc	C	RL	Rg 2	Rc	Cc	C	RL	Rg 2	Rc	Cc	C	
	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	
Rg 2	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	
Rc	2800	3400	3800	4800	6400	7500	11400	14500	17300	2200	2700	3100	3900	5300	6200	
Cc	2	1.62	1.3	1.12	0.84	0.66	0.52	0.4	0.33	2.2	2.1	1.85	1.7	1.25	1.2	
C	0.05	0.025	0.01	0.025	0.01	0.005	0.01	0.006	0.004	0.055	0.03	0.015	0.035	0.015	0.008	
Eo 3	14	17	20	16	22	23	18	23	26	34	45	54	41	54	59	
V.G. 4	9	9	10	10	11	12	12	12	13	10	11	11	12	13	13	

DUPLEX-DIODE TRIODE TYPE 6Q7

Ebb 1	90					180					300					Ebb 1
	RL	Rg 2	Rc	Cc	C	RL	Rg 2	Rc	Cc	C	RL	Rg 2	Rc	Cc	C	
	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	
Rg 2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	0.1	0.25	0.5	1	2	
Rc	4000	4200	4300	7200	7600	8000	11500	12300	13700	1600	1900	2100	3400	4000	4500	
Cc	2.07	1.7	1.5	1.17	1.2	0.9	0.72	0.6	0.45	3	2.5	2.3	1.6	1.3	1.05	
C	0.02	0.01	0.005	0.01	0.006	0.003	0.006	0.003	0.0015	0.02	0.01	0.005	0.01	0.005	0.003	
Eo 3	5	8	9	8	11	13	9	13	17	19	26	29	25	31	37	
V.G. 4	23	a	28	b	29	c	31	33	37	28	33	35	36	38	40	

DUPLEX-DIODE TRIODE TYPE 6R7

Ebb 1	90					180					300					Ebb 1
	RL	Rg 2	Rc	Cc	C	RL	Rg 2	Rc	Cc	C	RL	Rg 2	Rc	Cc	C	
	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	
Rg 2	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	
Rc	2300	2600	2900	3500	4400	5000	7600	9800	11300	1700	2100	2500	3000	4100	4600	
Cc	2	1.7	1.27	1.2	0.9	0.77	0.54	0.42	0.38	2.3	1.9	1.5	1.3	0.9	0.8	
C	0.05	0.03	0.01	0.03	0.01	0.006	0.015	0.007	0.003	0.05	0.03	0.01	0.03	0.01	0.006	
Eo 3	14	18	20	15	19	21	15	18	21	31	40	45	35	43	46	
V.G. 4	8	9	10	10	10	11	10	11	11	9	9	10	10	10	11	

DUPLEX-DIODE TRIODE TYPES: 55, 85

Ebb 1	90					180					300					Ebb 1
	RL	Rg 2	Rc	Cc	C	RL	Rg 2	Rc	Cc	C	RL	Rg 2	Rc	Cc	C	
	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	
Rg 2	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	0.05	0.1	0.25	0.5	1	
Rc	3800	4600	5400	6620	9000	10300	15100	20500	24400	3200	4100	5000	6200	8700	10000	
Cc	1.4	1.1	0.86	0.7	0.55	0.5	0.31	0.25	0.2	1.8	1.6	1.2	0.9	0.7	0.57	
C	0.06	0.03	0.015	0.04	0.015	0.007	0.015	0.007	0.004	0.06	0.045	0.02	0.04	0.015	0.008	
Eo 3	16	19	23	17	22	25	18	23	26	33	44	49	37	47	50	
V.G. 4	4.5	4.9	5.1	5.1	5.4	5.5	5.3	5.5	5.6	4.9	5.2	5.3	5.3	5.5	5.6	

1 Voltage at plate equals plate-supply voltage minus voltage drop in RL and Rc. For other supply voltages differing by as much as 50% from those listed, the values of resistors, grid bias, and gain are approximately correct. The value of voltage output, however, for any of these other supply voltages equals the listed voltage output multiplied by the new plate-supply voltage divided by the plate-supply voltage corresponding to the listed voltage output.

2 For following stage (see Circuit Diagrams).

3 Voltage across Rg at grid-current point.

4 Voltage Gain at 5 volts (RMS) output unless index letter indicates otherwise.

a At 2 volts (RMS) output.

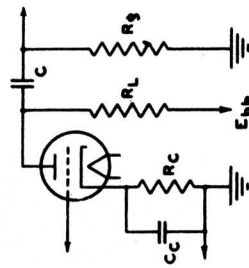
b At 3 volts (RMS) output.

c At 4 volts (RMS) output.

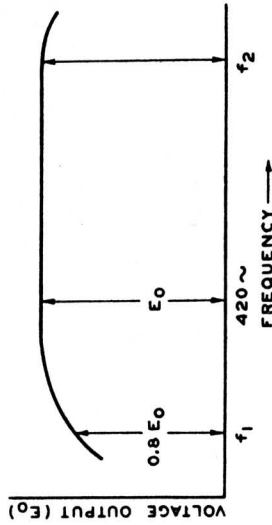
d At 2.2 volts (RMS) output.

See next page for Diagrams and Notes.

TRIODE DIAGRAM WITH LEGEND



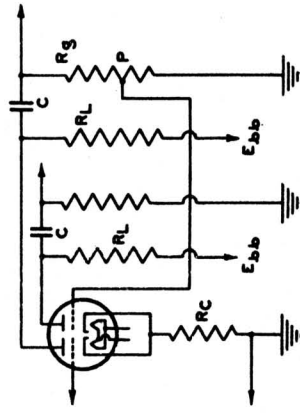
FREQUENCY CHARACTERISTIC OF SINGLE-STAGE RESISTANCE-COUPLED TRIODE AMPLIFIER



NOTES

- Condensers C and Cc have been chosen to give output voltages equal to 0.8 Eo for f₁ of 100 cycles. For any other value of f₁, multiply values of C and Cc by 100/f₁.
- In the case of condenser Cc, the values shown are for an amplifier with d-c heater excitation. When a-c is used, depending on the character of the associated circuits, the gain, and the value of f₁, it may be necessary to increase the value of Cc to minimize hum disturbances. It may also be desirable to have a d-c potential difference of approximately 10 volts between heater and cathode.
- f₂ = frequency at which high-frequency response begins to fall off.
- The voltage output at f₁ for π like stages equals (0.8 Eo)ⁿ.
- Decoupling filters are not necessary for two stages or less.
- For an amplifier of typical construction, the value of f₂ is well above the audio-frequency range for any value of RL.
- Always use highest permissible value of Rg.
- A variation of ± 10% in values of resistors and condensers has only a slight effect on performance.

TWIN-TRIODE DIAGRAM WITH LEGEND



FREQUENCY CHARACTERISTIC OF RESISTANCE-COUPLED TWIN-TRIODE AMPLIFIER

SEE INFORMATION UNDER TRIODE AMPLIFIER WHICH IN GENERAL APPLIES ALSO TO THIS CASE.

NOTES

- The diagram given above is for Phase-Inverter Service. The signal input is supplied to the grid of the left-hand triode unit. The grid of the right-hand unit obtains its signal from a tap (P) on the grid resistor (Rg) in the output circuit of the left-hand triode unit. The tap (P) is chosen so as to make the voltage output of the right-hand unit equal to that of the left-hand unit. Its location is determined from the voltage gain values given in the Chart. For example, if the value of voltage gain is 20 (from the Chart), (P) is chosen so as to supply 1/20 of the voltage across (Rg) to the grid of the right-hand triode.
- For phase-inverter service, the cathode resistor (Rc) should not be by-passed by a condenser. Omission of the condenser in this service assists in balancing the output voltages. The value of (Rc) is specified on the basis that both units are operating simultaneously at the same values of plate load and plate voltage.



RCA MANUFACTURING COMPANY, INC.

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Harrison, New Jersey

RCA RADIOTRON
D I V I S I O N

EXPLANATION OF TWIN-TRIODE R_o VALUES

APPLICATION NOTE No. 67
ON
RESISTANCE-COUPLED AUDIO-FREQUENCY AMPLIFIERS

In the Triode Chart of this Note, it should be observed, as stated in Notes under Twin-Triode Diagram, that the R_o values for Twin-Triode types 6A6, 6N7, 53, and 79 are for two triode units on basis that tube type is used in phase-inverter service. Other tabulated data on these types are given for each triode unit.